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Patentanmeldung Nr. Patent application No. Demande de brevet n°

99410173.1

Der Präsident des Europäischen Patentamts;
Im Auftrag

For the President of the European Patent Office

Le Président de l'Office européen des brevets
p.o.

I.L.C. HATTEN-HECKMAN

DEN HAAG, DEN
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**Blatt 2 der Bescheinigung
Sheet 2 of the certificate
Page 2 de l'attestation**

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DSL TRANSMISSION SYSTEM WITH FAR-END CROSSTALK COMPENSATION**FIELD OF THE INVENTION**

The present invention relates to digital subscriber line transmission systems, which allow, in particular high speed communication on twisted pair telephone lines based on discrete
5 multitone transmission (DMT). The invention relates more specifically to a far-end crosstalk (FEXT) canceller for compensating the crosstalk signal induced by modems located at the far-end of such a transmission system.

DISCUSSION OF THE RELATED ART

10 Figure 1 schematically shows a modem, comprising a transmission (TX) and a reception (RX) section in a conventional DSL transmission system using digital multitone. A serial stream of data X is provided to a mapper circuit 11 mapping each data
15 into a symbol of a constellation, for example of a QAM (Quadrature Amplitude Modulation) constellation. The mapped values are then transformed into a set S of N components by a serial to parallel converter 12, each component of the set being considered as a frequency domain coefficient. This set of frequency domain coefficients, hereafter also called DMT symbol, is
20 provided to an inverse fast Fourier transform (IFFT) circuit 13 which generates a time domain block of samples and is followed by a parallel/serial converter (P/S). This time domain block is therefore the sum of N sinusoidal subcarriers of different frequencies, the amplitude thereof being determined by the

corresponding frequency domain coefficient received by the IFFT circuit.

Each time domain block is cyclically prefixed (cp) and suffixed (cs) in a block 19 to suppress or at least attenuate the Inter Symbol Interference (ISI) and the Inter Carrier Interference (ICI) caused by the channel, and is transmitted onto a telephone line 10 through a hybrid line interface 18. The line interface 18 also receives incoming time domain blocks from another modem connected to line 10.

At the receiving side, the incoming time domain blocks from line 10 are provided to a fast Fourier transform (FFT) circuit 14 through a block 19' that deletes the prefix and suffix and a serial/parallel converter (S/P) which calculates the N frequency domain coefficients for each block. The N frequency domain coefficients are then provided to an equalizer 15 which compensates for the attenuation and phase shift incurred by each frequency component. The equalized values are then serialized by a parallel to serial converter 16 into a stream of N complex numbers $R(f_j)$ and then processed by a demapper 17 attributing for each $R(f_j)$ the symbol \hat{S}_c of the constellation which comes closest thereto. The demapper 17 further outputs the digital word \hat{X}_c associated with the selected constellation point \hat{S}_c .

Figure 2 schematically shows a DSL transmission system comprising a central office 20 communicating with a plurality of end-users over telephone lines 25, 26. The modems M1, M2, Mc(1), Mc(2) have the structure represented in figure 1. The end of a telephone line connected to a modem of the central office is called the line termination (LT) side while the end connected to a modem of an end-user is called the network termination (NT) side.

Ideally, such a DSL transmission system allows the whole frequency band to be used for simultaneous full-duplex transmissions. However, in practice, different sources of noise disturb the transmissions and impede proper reception of data.

For a given modem, three different sources of noise can be distinguished as illustrated on figure 2:

- the self-echo, i.e. for a given modem, the parasitic signal from the transmitter TX leaking to the receiver RX through the hybrid interface;

- the near-end crosstalk (NEXT) arising from signals in adjacent telephone lines 25, 26 with opposite transmission directions. More specifically, in the present example, the NEXT generated at the modem Mc(1) is the parasitic signal received by this modem from the modem Mc(2). In this instance the NEXT is called NT-NEXT because the modem Mc(1) is located on the NT side. Reciprocally, the NEXT generated at modem M1 by the modem M2 is called LT-NEXT;

- the far-end crosstalk (FEXT) arises from signals traveling along the same transmission direction in adjacent telephone lines. More precisely, in the illustrated example, the FEXT generated at the modem Mc(1) is the parasitic signal received by this modem from the modem M2 located on the opposite side, due to the coupling between the telephone lines 25 and 26 sharing a common binder. In this instance the FEXT is called NT-FEXT because the modem Mc(1) is located on the NT side. Reciprocally, the FEXT generated at modem M1 by the modem Mc(2) is called LT-FEXT.

Echo cancellers for canceling self-echoes are known, e.g. from unpublished European patent application N° 98410112 filed by the applicants. This document is however to be taken into account as prior art only under Art. 54(3) EPC.

There is also known from US-A-5887032 a canceller for canceling out the NEXT interference in an ADSL transmission system on the LT side. This canceller operates in the frequency domain and assumes, for a given subcarrier or tone, that the NEXT interference is proportional to the symbol value emitted by the modem transmitting on the interfering channel. The latter value is scaled by a given coefficient and subtracted from the symbol received by the modem suffering from the NEXT interference.

Both self-echo cancellation and LT-NEXT cancellation are possible because the signal transmitted by the same modem (in the case of the self-echo) or by a neighboring modem of the

central office (in the case of LT-NEXT interference) is directly available.

FEXT cancellation is however intrinsically more complex than NEXT or self-echo cancellation because the modem transmitting over the interfering channel is now located on the far-end side and the actual values of the interfering symbols are therefore not known.

SUMMARY OF THE INVENTION

An object of the present invention is therefore to design a canceller circuit for a DMT based DSL transmission system capable of significantly removing the FEXT interference and having a simple structure.

It is also an object of the present invention to design an efficient FEXT canceling method in a DMT based DSL transmission system.

These objects are attained by the invention as claimed.

The foregoing and other objects, features, aspects and advantages of the invention will become apparent from the following detailed description of embodiments, given by way of illustration and not of limitation with reference to the accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1, previously described, schematically shows the structure of a modem suitable for use in a DSL transmission system;

Figure 2, previously described, schematically shows the different types of noise occurring in a DSL transmission system;

Figure 3A shows a first embodiment of a FEXT canceller according to the invention;

Figure 3B shows a second embodiment of a first canceller according to the invention;

Figure 4 shows the structure of a modem on the LT side suitable for use with a FEXT canceller according to the second embodiment of the invention;

Figure 5 shows the structure of a modem on the NT side suitable for use with a FEXT canceller according to the second embodiment of the invention; and

Figure 6 shows the overall structure of a DSL transmission system using both an LT-FEXT canceller and an NT-FEXT canceller according to the first or the second embodiment of the invention.

DESCRIPTION OF THE INVENTION

The invention is based on the idea that it is possible to remove FEXT interference at the NT side by appropriately pre-distorting at the LT side the DMT symbols to be transmitted. More specifically, if S is the vector (S_i) , $i = 1$ to n , where S_i is the DMT symbol to be transmitted by the modem M_i and if R is the vector (R_i) , $i = 1$ to n , where R_i is the DMT symbol received by the modem $M_c(i)$, then $R = H \cdot S$ where H is the transfer matrix of the n downstream (i.e. LT to NT) transmission channels.

R and S are vectors of $n \cdot N$ components as concatenation of n vectors of N components, each S_i (resp. R_i) being a vector of N frequency (or tone) components $S_i(f_j)$ (resp. $R_i(f_j)$).

If we assume that there exists a matrix M such that $H \cdot M = D$ where D is diagonal and if the vector S is multiplied by the matrix M before transmission then $R = H \cdot M \cdot S = D \cdot S$ is freed from FEXT interference since any component $R_i(f_j)$ of the DMT symbol R_i received by the modem $M_c(i)$ depends only upon the component $S_i(f_j)$ of the DMT symbol S_i transmitted by the modem M_i .

In most practical cases, the transfer matrix H can be inverted and one can simply choose $M = H^{-1}$ and $D = I$, where I is unity matrix. This provides a further advantage since equalization of the frequency components of the received DMT symbols is no longer necessary.

A FEXT precompensating circuit according to a first embodiment of the invention is hereafter described in connection with figure 3A. This circuit comprises a memory 31 in which are stored the columns H_i of the transfer matrix H . The matrix coefficients H_{ik} are then supplied to a matrix inverter 32. A matrix multiplier 33 receives the inverted matrix $M = H^{-1}$ as well as the

DMT symbols S_i provided by the LT modems M_i and performs the multiplication M^*S . The DMT symbols

$$(M^*S)_i = (M^*S)_{j+(i-1)*N}, \quad j = 1 \text{ to } N,$$

are then respectively supplied to the LT modems M_i . In each LT
 5 modem M_i the DMT symbol $(M^*S)_i$ is provided to an IFFT circuit in place of the DMT symbol S_i and the resulting time block of samples is transmitted to the NT modem $M_c(i)$.

As shown in figure 6, the coefficients of the transfer
 matrix can be provided by an external LT-FEXT canceller 60 such
 10 as disclosed in copending European application entitled "DSL transmission system with far-end crosstalk cancellation" filed by the applicants. Basically, this LT-FEXT canceller adaptively estimates the inverse H_u^{-1} of the transfer matrix of the n upstream (i.e. NT to LT) transmission channels and multiplies the
 15 vector of the DMT symbols received by the LT modems by H_u^{-1} for recovering FEXT-free received symbols. Assuming that the downstream transfer matrix $H = H_u^{-1}$, the FEXT precompensating circuit 61 can directly use the estimated matrix output by the LT-FEXT canceller as matrix H .

20 When there is no symmetry, one can also estimate H^{-1} or H , respectively $H(fj)^{-1}$ or $H(fj)$ as mentioned hereafter, by an adaptative algorithm minimizing a quadratic error (for example by a least mean square method). The method consists in calculating the error

$$25 \quad e_k^2 = \|R_k - \hat{S}_k\|^2 = \|(H \cdot \hat{H}^{-1} \cdot S)_k - \hat{S}_k\|^2$$

on the NT side and in sending back this error to the LT side for updating the coefficients of \hat{H}^{-1} (here, k is not the time index but a specific value of S).

A second embodiment of the FEXT precompensating circuit
 30 is hereafter described in connection with figure 3B.

In a conventional DMT transmission system, the FEXT generated at a frequency fj depends also upon frequency components transmitted at neighboring frequencies since the limited duration of the time domain blocks causes a spreading of the frequency components. It is now assumed that the FEXT generated at a
 35

frequency f_j depends only upon components transmitted at the same frequency (intra-frequency FEXT).

If the modems are synchronous Zipper modems as described in the international application W097/06619, the assumption made above is fully valid, i.e. the FEXT at a frequency f_j is independent from the frequency components transmitted at the frequencies f_i , $i \neq j$. Indeed, in such modems, the suffix appended to each time domain block before transmission eliminates any inter-frequency crosstalk.

With the intra-frequency crosstalk assumption the matrix calculations are simplified: if the components at each frequency f_j of the DMT symbols S_i (resp. R_i) in vector S (resp. R) are grouped together, the transfer matrix H exhibits the form of a diagonal block matrix with N matrices $H(f_j)$ along its diagonal, the intra-frequency crosstalk $H(f_j)$ matrices having $n \times n$ coefficients. The inversion of the matrix H and the multiplication $M \cdot S$ can thus be performed sequentially frequency by frequency.

In this embodiment a sequence controller 34' controls an address generator 35', parallel to serial converters 36-1' to 36-n', serial to parallel converters 37-1' to 37-n', and generates the insertion control clock CK_{in} and the tone clock CK_t for all the LT modems i .

The memory 31' is organized in planes, each plane storing a matrix $H(f_j)$. The matrices $H(f_j)$ are sequentially retrieved and inverted in matrix inverter 32'. The inverted matrices $H^{-1}(f_j)$ are sequentially provided to matrix multiplier 33'.

The DMT symbols S_i are each provided to a parallel to serial converter 36-i' controlled by the sequence controller 34'. The N components $S_i(f_j)$, $j = 1$ to N , are sequentially output by the parallel to serial converters and the matrix multiplier 33' sequentially calculates the matrix products $H^{-1}(f_j) \cdot S(f_j)$, where $S(f_j)$ is the vector $S(f_j) = (S_i(f_j))$, $i = 1$ to n . The components $[H^{-1}(f_j) \cdot S(f_j)]_i$ of the vector provided by the matrix multiplier 33' are then each transformed by serial to parallel converters 36-i' into DMT symbols $(H^{-1} \cdot S)_i$.

Figure 4 shows an LT-modem M_i for use with a FEXT precompensating circuit according to the second embodiment of the invention.

The structure of this modem is similar to the one shown in figure 1, the identical elements bearing translated references. After mapping, the complex components are provided to a zero inserter inserting a zero component on a rising edge of signal CK_{in} at a location given by a modulo N counter 422 clocked by the signal CK_t . The complex components are then converted into a DMT symbol S_i by a serial to parallel converter 412 and S_i is provided to a precompensating circuit 30'. The DMT symbol $(H^{-1} * S)_i$ output by the precompensating circuit is provided to a pilot tone inserter circuit 430 which inserts a pilot tone component $P(i, f_j)$ on a rising edge of the signal CK_{in} delayed by the delay 431. The modulo N counter 422 provides the more significant bit (MSB) of a read address to ROM 433, the least significant bit (LSB) of the address being given by a signal SEL_i which indicates whether the modem is selected. This signal can be directly generated by the sequencer or, preferably, provided by a comparator 434 comparing the rank i of the modem M_i with the output of a modulo n counter 432 clocked by CK_{in} . The data $P(i, f_j)$ stored at the read address are supplied to the pilot tone inserter 430 which inserts the pilot tone value $P(i, f_j)$ as a component at frequency component f_j of the DMT symbol S_i on the rising edge of the delayed signal CK_{in} . The DMT symbol is then frequency-time transformed by the IFFT circuit 413 as usual.

On the receiving side of the modem M_i , the block RX is identical to the block RX of the modem illustrated in figure 1. The data output by RX are provided to a demultiplexer which separates the received coefficients H_{ik} from the data as explained further below.

In normal transmitting mode the signals CK_{in} and CK_t are low, no zero is inserted by the zero inserter 420 and no pilot tone value is inserted by the inserter 430. The data X are normally mapped and the parallelized into DMT symbols S_i . The

precompensated DMT symbol $(H^{-1} \cdot S)_i$ received from the precompensating circuit 30' is also directly fed to the IFFT circuit.

In mixed transmitting/updating mode, the zero inserter inserts a zero complex value at the location j given by the counter 422. This is done simultaneously for all the modems M_i . Each S_i carries therefore $N-1$ normally mapped data and a zero at the frequency f_j . All the precompensated DMT symbols $(H^{-1} \cdot S)_i$ received from the precompensating circuit 30' have therefore also a zero as component at frequency f_j . In each modem M_i , this zero is replaced by the pilot tone inserter with a complex value $P(i, f_j) = 0$ if $SEL_i = 0$ and $P(i, f_j) = p_j$ if $SEL_i = 1$, where p_j is a predetermined complex number depending upon the frequency f_j only. The modems $M_1, M_2 \dots M_n$ are selected in turn at the frequency of CK_{in} . A further CK_t pulse increments j and the whole insertion process is repeated again at the new frequency f_j .

If the current counter value is j_0 and if the modem M_{i_0} is selected, the n components at frequency f_{j_0} of the DMT symbols received by the NT modems $M_c(i)$ constitute the i_0 -th column of the matrix $H(f_{j_0})$ multiplied by p_{j_0} . The NT modems $M_c(i)$ can therefore transmit the matrix coefficients $H_{i_0 i}(f_{j_0})$ to the LT modems M_i , each coefficient being for example preceded by a predetermined header. Demultiplexer 440 in LT modem M_i detects this header and extracts the coefficient $H_{i_0 i}(f_{j_0})$. The column vector $(H_{i_0 i}(f_{j_0}))$, $i = 1$ to n , is sent to the precompensating means 30' and stored at the i_0 -th column of the j_0 -th plane of memory 31'.

Figure 5 shows the structure of a NT modem which can be used with the precompensating circuit of figure 3B and LT modems of figure 4. The structure of this modem is similar to the one shown in figure 1, the identical elements bearing translated references. In contrast with the latter, the NT modem comprises a multiplexer 520, updating means 530, a modulo N counter 522 and has no equalizer. It is assumed that the NT side is provided with a sequence controller synchronized with the controller 34' and having the same sequence pattern.

The updating means 530 receives a value j_0 from the modulo N counter 522 clocked by CK_t and the signal CK_{in} delayed

by delay 531. In updating/transmitting mode, the updating means 530 extracts on a rising edge of C_{kin} the component at frequency j_0 of the received DMT symbol output by the FFT block 514. The updating means 530 compares the coefficient $H_{i0i}(fj_0)$ with the last coefficient $H_{i0i}(fj_0)$ previously obtained. If the absolute value of the difference is greater than a given threshold Thj_0 , which in general is a function of frequency fj_0 , the new coefficient is appended to a header and transmitted via multiplexer 520 to the LT modem M_i .

Turning back to figure 3B, similarly to the first embodiment, the columns of the matrices $H(fj)$ can also be directly provided by an external LT-FEXT canceller as shown in figure 6. With the assumption of intra-frequency FEXT the LT-FEXT canceller has a simplified structure. In such an instance, as disclosed in the above cited copending application, the LT-FEXT canceller estimates the matrices $H_u^{-1}(fj)$ for $j = 1$ to N . These matrices can be used as the precompensating matrices $H(fj)$.

Such alterations, modifications, and improvements are intended to be part of this disclosure, and are intended to be within the spirit and the scope of the present invention. Accordingly, the foregoing description is by way of example only and is not intended to be limiting. The present invention is limited only as defined in the following claims and the equivalents thereto.

CLAIMS

1. A far-end crosstalk (FEXT) canceling circuit for a digital subscriber line transmission system, said transmission system comprising a plurality n of line termination (LT) modems (M_i) transmitting discrete multitone (DMT) symbols (S_i) to corresponding network (NT) termination modems ($Mc(i)$) over n transmission channels, characterized in that it comprises precompensation means (33, 33') multiplying, before transmission, the vector $S = (S_i)$, $i = 1$ to n , by a precompensation matrix (M) such that the matrix product $H*M$ is diagonal, H being the transfer matrix of the n transmission channels defined by $R = H*S$, where $R = (R_i)$, $i = 1$ to n , is the vector of the DMT symbols R_i respectively received by the modems $Mc(i)$.

2. The FEXT canceling circuit of claim 1, characterized in that it further comprises:
storing means (31) storing said transfer matrix (H);
inversion means (32) inverting said transfer matrix and providing the precompensation means (33) with the inverted matrix (H^{-1}).

3. The FEXT canceling circuit of claim 1, characterized in that it further comprises:

storing means (31') storing transfer matrices $H(f_j)$ of the n transmission channels at tones f_j , $H(f_j)$ being defined by $R(f_j) = H(f_j) * S(f_j)$, where $R(f_j)$ is the vector $R(f_j) = (R_i(f_j))$, $i = 1$ to n and $S(f_j)$ is the vector $S(f_j) = (S_i(f_j))$, $i = 1$ to n , $R_i(f_j)$ and $S_i(f_j)$ being the components at tone f_j of the received DMT symbol R_i and transmitted DMT symbol S_i respectively; and
inversion means (32') sequentially inverting said transfer matrices $H(f_j)$ and supplying the precompensation means (33') with the inverted matrices $H^{-1}(f_j)$, the precompensation means (33') sequentially calculates the products $H^{-1}(f_j) * S(f_j)$.

4. A digital subscriber line transmission system comprising a plurality n of line termination (LT) modems (M_i) transmitting discrete multitone (DMT) symbols S_i to corresponding

network (NT) termination modems ($M_c(i)$) over n transmission channels, characterized in that it further comprises:

a FEXT canceling circuit (61) according to claim 1 or 2 canceling FEXT at the NT side of said system; and

5 an LT-FEXT canceling circuit (60) canceling FEXT at the LT side of said system by estimating the inverse of the transfer matrix H^{-1}_{up} of the plurality of the transmission channels in the upstream direction, said LT-FEXT canceling circuit supplying the storing means of said FEXT canceling circuit with $H = H^{-1}_{up}$.

10 5. A digital subscriber line transmission system comprising a plurality n of line termination (LT) modems (M_i) transmitting discrete multitone (DMT) symbols S_i to corresponding network (NT) termination modems ($M_c(i)$) over n transmission channels, characterized in that it further comprises:

15 a FEXT canceling circuit (61) according to claim 3 canceling FEXT at the NT side of said system; and

an LT-FEXT canceling circuit (60) canceling FEXT at the LT side of said system by estimating the inverse of the transfer matrices $H^{-1}_{up}(f_j)$ of the plurality of transmission channels in
20 the upstream direction at tone f_j , said LT-FEXT canceling circuit supplying the storing means of said FEXT canceling circuit with $H(f_j) = H^{-1}_{up}(f_j)$.

6. A digital subscriber line transmission system comprising a plurality n of line termination (LT) modems (M_i)
25 transmitting discrete multitone (DMT) symbols S_i to corresponding network (NT) termination modems ($M_c(i)$) over n transmission channels, further comprising a FEXT canceling circuit according to claim 3 canceling FEXT at the NT side of said system;

characterized in that

30 each LT modem (M_i) comprises inserting means (430) for inserting at predetermined times known symbols $P(i, f_j)$ as components at tone f_j of multitone symbols S_i ;

each NT modem $M_c(k)$ comprises means (530) for detecting at said predetermined times the components $R_k(f_j)$ and for deriving
35 therefrom the coefficients $H_{ik}(f_j)$ of the transfer matrix $H(f_j)$;

each NT modem $M_c(k)$ further comprises a multiplexer (520) for multiplexing data to be transmitted with said coefficients $H_{ik}(f_j)$;

each LT modem M_k further comprises a demultiplexer (440) for extracting from the received data said coefficients $H_{ik}(f_j)$.

7. A far-end crosstalk (FEXT) canceling method for a digital subscriber line transmission system, said transmission system comprising a plurality n of line termination (LT) modems (M_i) transmitting discrete multitone (DMT) symbols S_i to corresponding network (NT) termination modems ($M_c(i)$) over n transmission channels, characterized in that the vector $S = (S_i)$, $i = 1$ to n , is multiplied, before transmission, by a precompensation matrix M such that the matrix product $H*M$ is diagonal, H being the transfer matrix of the n transmission channels defined by $R = H*S$, where $R = (R_i)$, $i = 1$ to n , is the vector of the DMT symbols R_i respectively received by the modems $M_c(i)$.

8. The FEXT canceling method of claim 7, characterized in that

said transfer matrix is stored in storing means;

said transfer matrix is then retrieved and inverted;

and

the inverted matrix is used as precompensation matrix M .

9. The FEXT canceling method of claim 7, characterized in that

the transfer matrices $H(f_j)$ of the n transmission channels at tones f_j are stored in storing means, $H(f_j)$ being defined by $R(f_j) = H(f_j)*S(f_j)$ where $R(f_j)$ is the vector $R(f_j) = (R_i(f_j))$, $i = 1$ to n , and $S(f_j)$ is the vector $S(f_j) = (S_i(f_j))$, $i = 1$ to n , $R_i(f_j)$ and $S_i(f_j)$ being the components at tone f_j of the received DMT symbol R_i and transmitted DMT symbol S_i respectively;

said transfer matrices $H(f_j)$ are retrieved and

inverted;

the inverted matrices $H^{-1}(f_j)$ are used as precompensating matrices at tones f_j .

10. The FEXT canceling method of claim 9, characterized in that

5 for each LT modem M_i , known symbols $P(i, f_j)$ are inserted at predetermined times as components at tone f_j of multitone symbols S_i ;

10 for each NT modem $M_c(k)$, the components $R_k(f_j)$ are detected at said predetermined times, the coefficients $H_{ik}(f_j)$ of transfer matrix $H(f_j)$ are derived therefrom and multiplexed with the data to be transmitted;

 for each modem M_k , said coefficients $H_{ik}(f_j)$ are extracted from the received data.

DSL TRANSMISSION SYSTEM WITH FAR-END CROSSTALK COMPENSATION

Abstract

The invention relates to a far-end crosstalk (FEXT) canceling circuit for a digital subscriber line transmission system, said transmission system comprising a plurality n of line termination (LT) modems (M_i) transmitting discrete multitone (DMT) symbols S_i to corresponding network (NT) termination modems ($M_c(i)$) over n transmission channels.

The invention proposes to multiply the vector $S = (S_i)$ $i = 1$ to n , before transmission, by a precompensation matrix M such that the matrix product $H \cdot M$ is diagonal, H being the transfer matrix of the n downstream transmission channels defined by $R = H \cdot S$ where $R = (R_i)$, $i = 1$ to n , is the vector of the DMT symbols R_i respectively received by the modems $M_c(i)$.

(Figure 3)

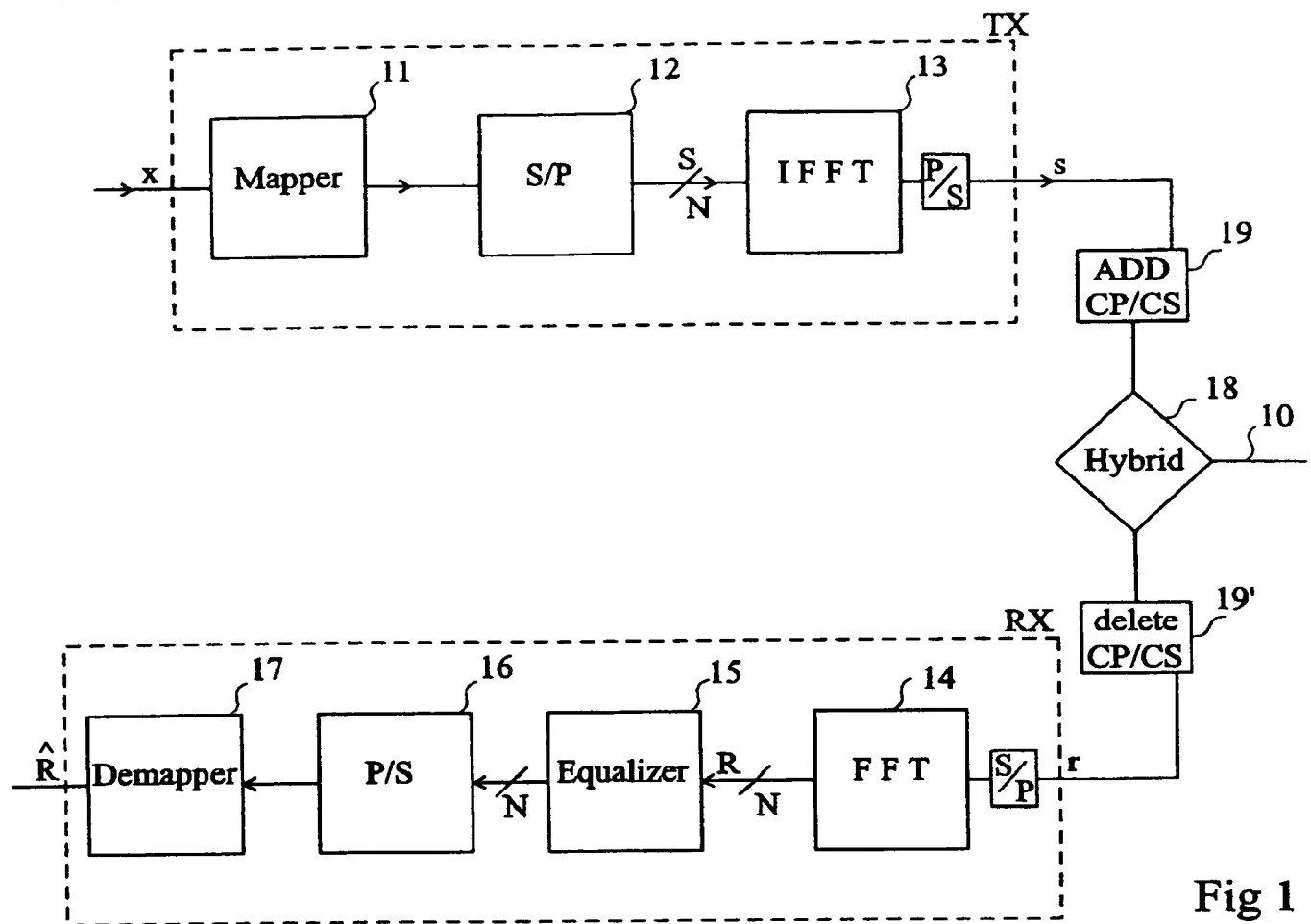


Fig 1

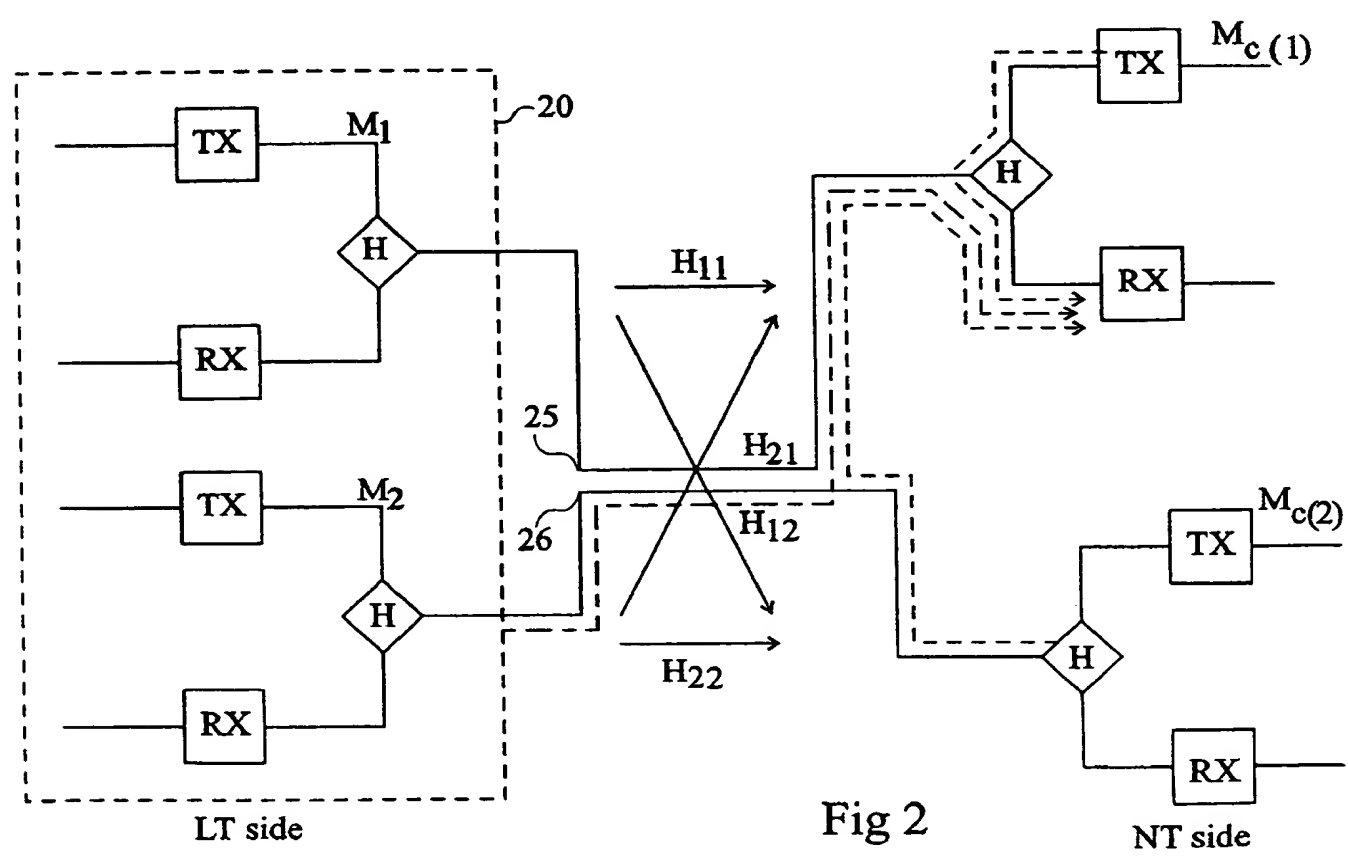


Fig 2

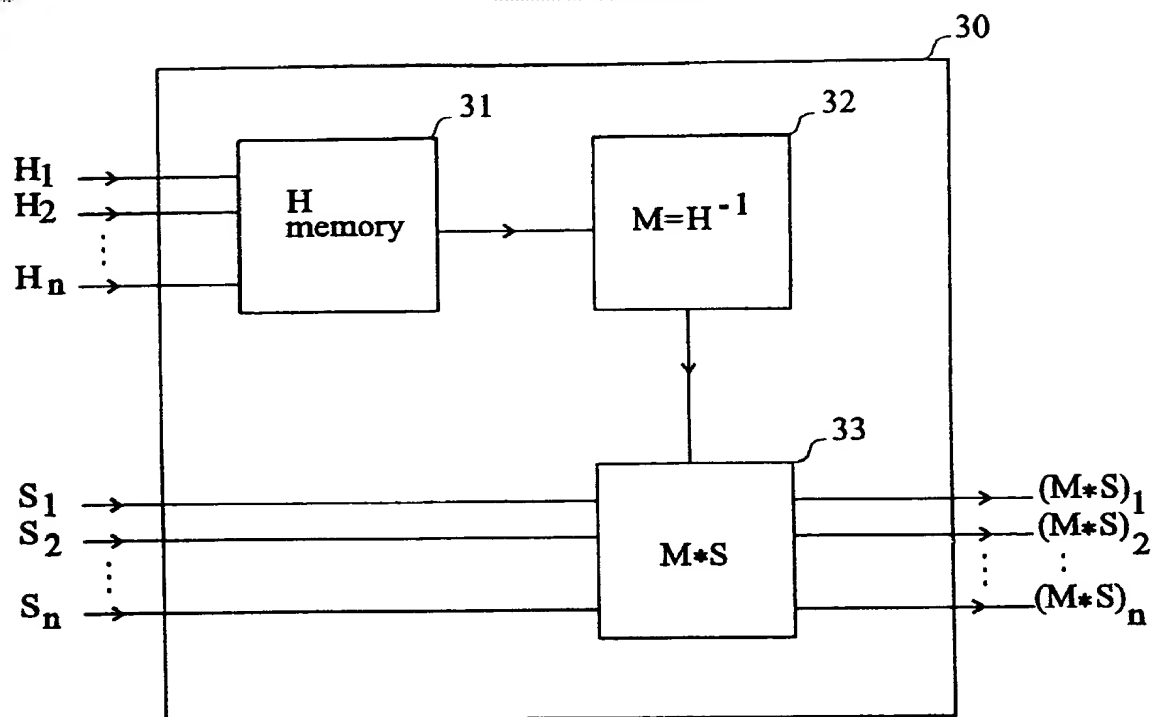


Fig 3A

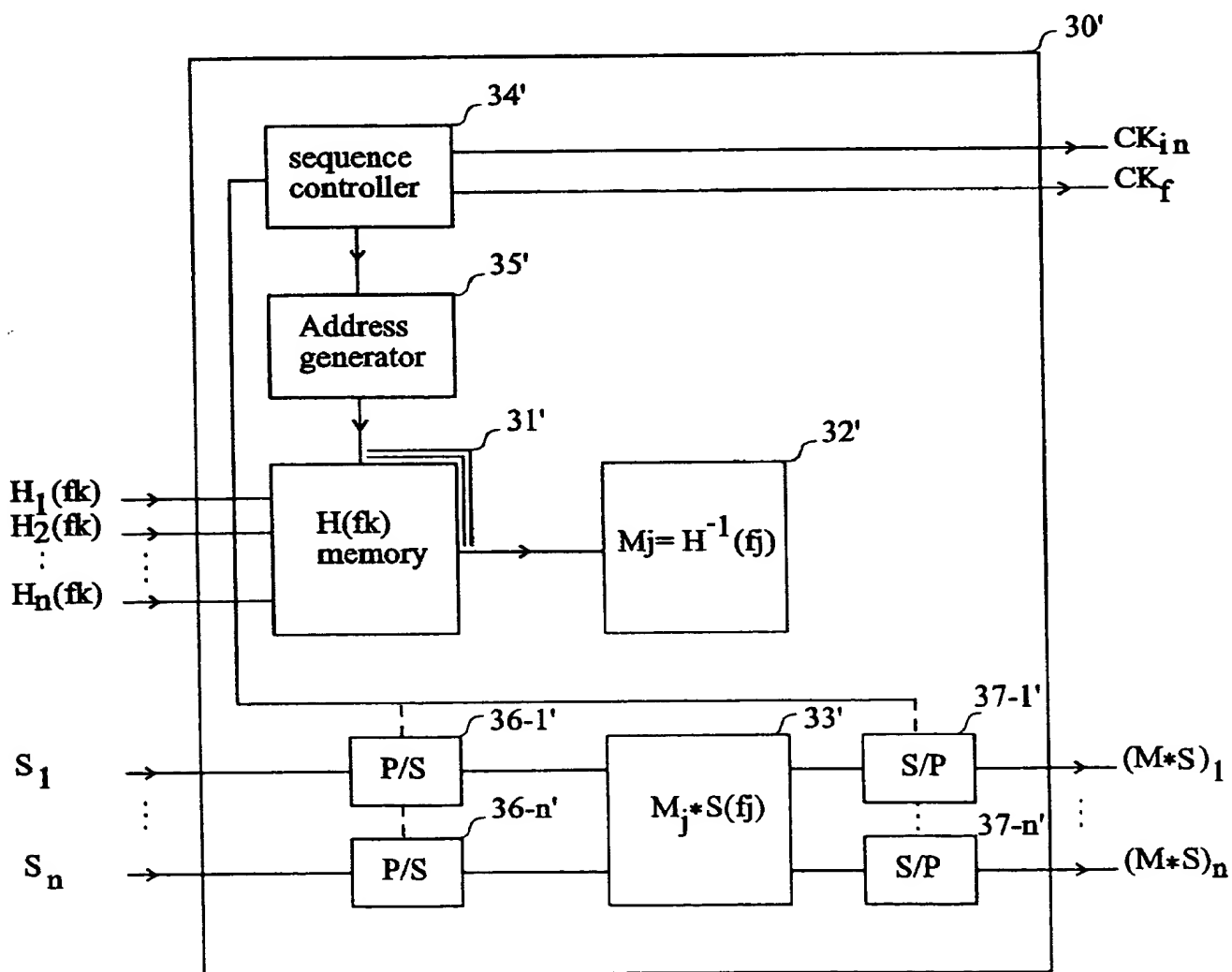
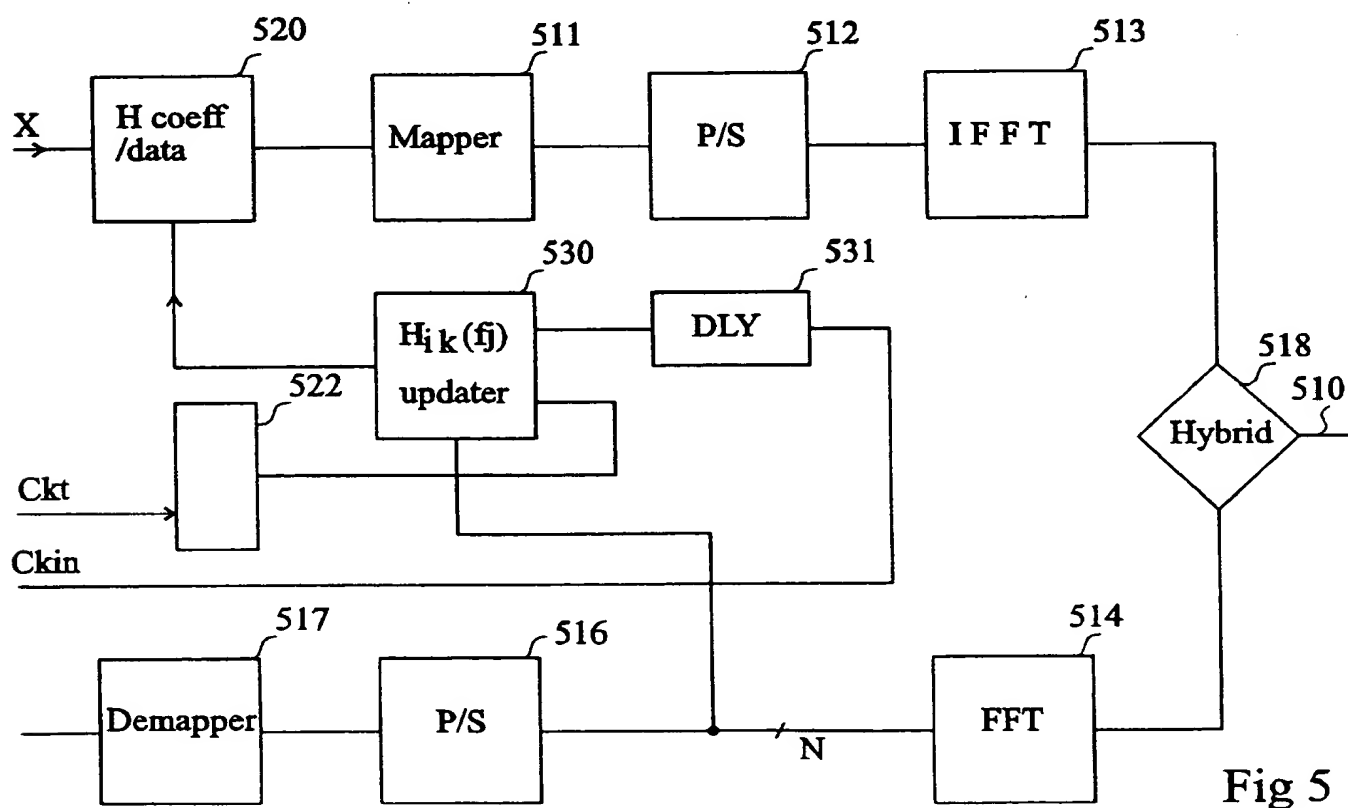
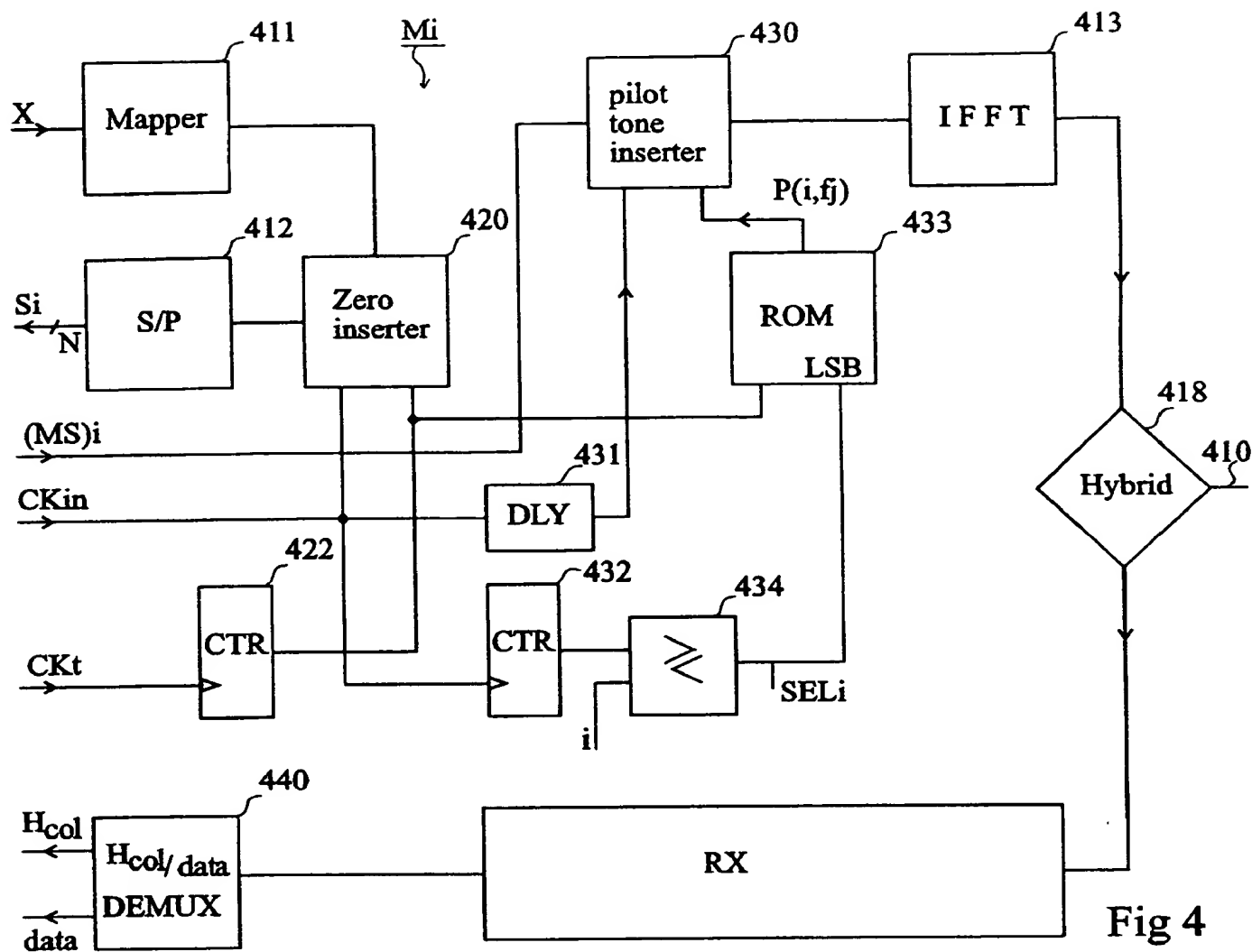


Fig 3B



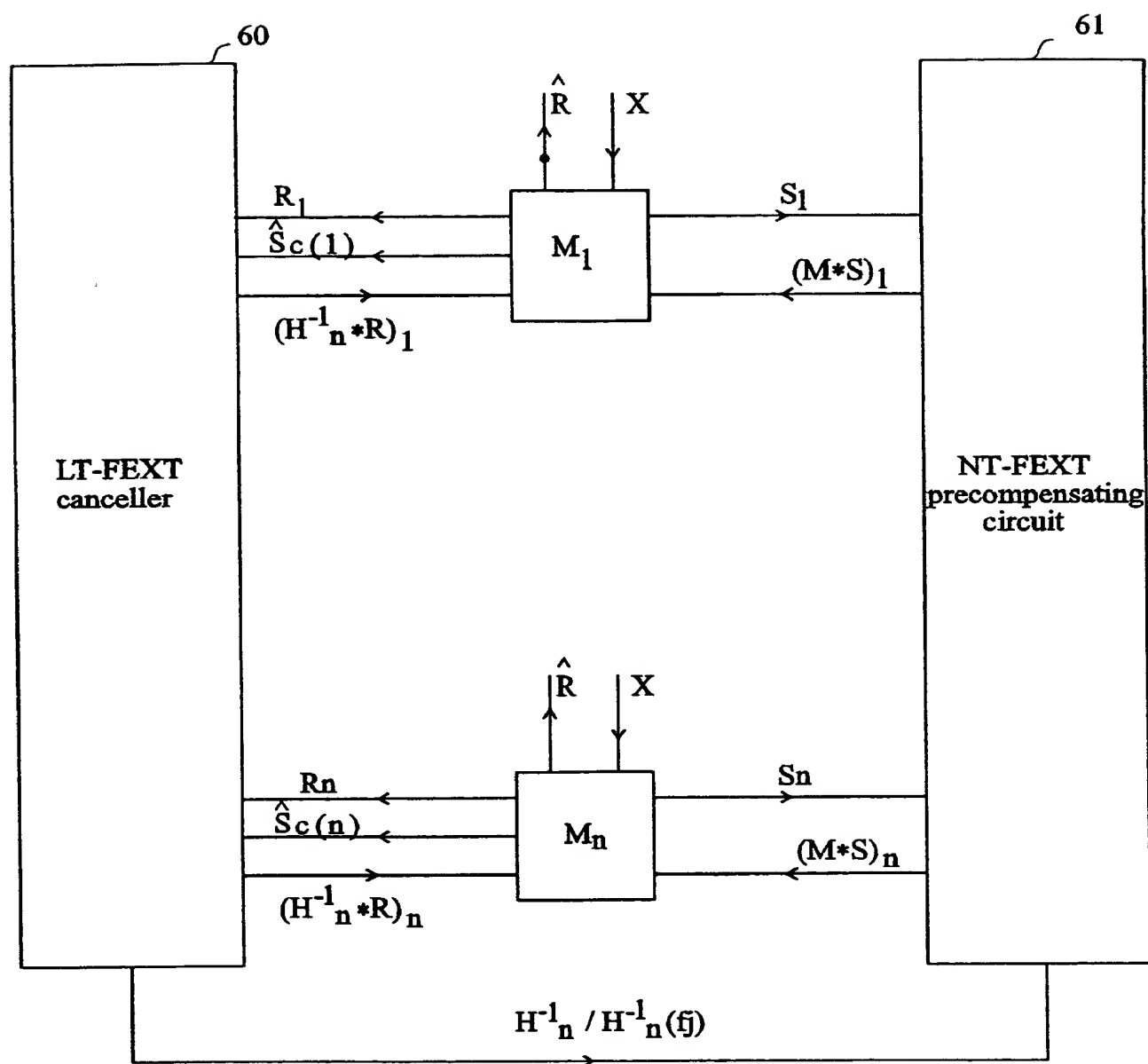


Fig 6